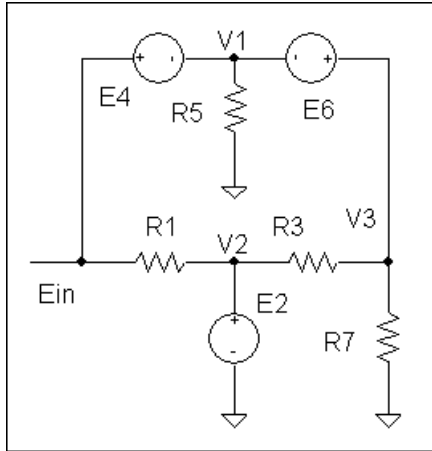


### DS Method – Twin-T Network (60 Hz Notch Filter)

The converted circuit with 1V sources replacing the capacitors is



By noting that  $V2 = E2 = 1V$ , one unknown node is eliminated. Hence  $N = 3$ ,  $U = 2$ ,  $M = 1$ . There are now two unknown nodes,  $V1$  &  $V3$ . The output is taken from the second unknown node  $V3$ , hence  $Y = 2$ .

Component values are given in the M-file following. We need  $U+N = 5$  dc equations. At node  $V1$  we use KCL to get

$$i_{C4} + i_{C6} = \frac{V1}{R5}, \quad \text{or} \quad \frac{V1}{R5} - i_{C4} - i_{C6} = 0 \quad \text{placing unknowns on LH side.}$$

At node  $V2 = E2$ , KCL gives

$$\frac{E_{in} - E2}{R1} = i_{C2} + \frac{E2 - V3}{R3} \quad \text{or} \quad i_{C2} - \frac{V3}{R3} = \frac{E_{in}}{R1} - E2 \left( \frac{1}{R1} + \frac{1}{R3} \right) \quad \text{knowns on RH side.}$$

At node  $V3$ , KCL gives

$$\frac{E2 - V3}{R3} = \frac{V3}{R7} + i_{C6} \quad \text{or} \quad V3 \left( \frac{1}{R3} + \frac{1}{R7} \right) + i_{C6} = \frac{E2}{R3}$$

Using KVL for the two capacitors ( $E4$  and  $E6$ )

$$V1 = E_{in} - E4, \quad V3 - V1 = E6$$

The  $B1$  array is skipped this time. The  $A1$  array below contains the coefficients of the LH sides of the foregoing equations, and the  $B2$  array contains the coefficients of the RH sides.

$$\begin{array}{ccccc}
 V1 & V3 & iC2 & iC4 & iC6 \\
 A1 = \begin{bmatrix} \frac{1}{R5} & 0 & 0 & -1 & -1 \\ 0 & \frac{-1}{R3} & 1 & 0 & 0 \\ 0 & \frac{1}{R3} + \frac{1}{R7} & 0 & 0 & 1 \\ 1 & 0 & 0 & 0 & 0 \\ -1 & 1 & 0 & 0 & 0 \end{bmatrix}
 \end{array}
 \quad
 \begin{array}{cccc}
 E2 & E4 & E6 & Ein \\
 B2 = \begin{bmatrix} 0 & 0 & 0 & 0 \\ -E2\left(\frac{1}{R1} + \frac{1}{R3}\right) & 0 & 0 & \frac{Ein}{R1} \\ \frac{E2}{R3} & 0 & 0 & 0 \\ 0 & -E4 & 0 & Ein \\ 0 & 0 & E6 & 0 \end{bmatrix}
 \end{array}$$

Note again that every row and column of A1 and every column of B2 must have at least one non-zero entry.

Following the sequence of E2, E4, and E6 in B2 above,

$$P = \begin{bmatrix} C2 & 0 & 0 \\ 0 & C4 & 0 \\ 0 & 0 & C6 \end{bmatrix}$$

Again the rest uses template equations as shown in the M-file TwinT.m.

### Introduction to Transient Analysis

Transient or time domain analysis is based on  $\frac{\Delta x}{\Delta t} = Ax + Bu$  or  $\Delta x = (Ax + Bu)\Delta t$

We already have the A & B arrays and we need to find the correct value of  $\Delta t$ . If  $\Delta t$  is too large the solution will not converge. If too small, excessive execution time may be required.

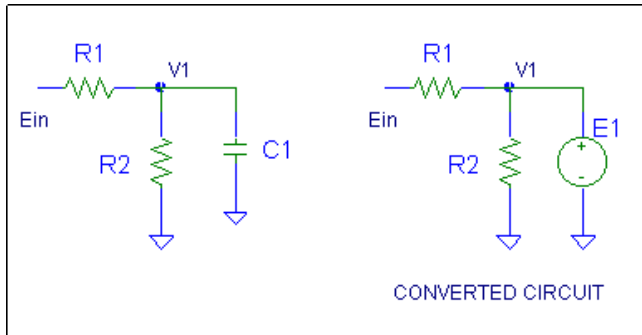
The time increment  $\Delta t$  generally should be smaller than the shortest time constant in A. Although this is not strictly true for every circuit and should only be used as a guide in selecting  $\Delta t$ . In MATLAB parlance

$$T_x = 1/\max(\max(\text{abs}(A))) \quad \text{Shortest time constant in matrix A.}$$

The time increment  $\Delta t$  is initially set at  $T_x/10$  and then the mantissa is rounded down to an integer value. For example,  $T_x = 8.99e-4$ , then  $\Delta t = 8e-5 = 80\mu s$ . Some adjustment of  $\Delta t$  may be required for circuits with very fast or very slow time constants.

Background information for the Time Response section of the M-file TwinT.m is given below:

A simple 1st order RC network is used as an example.



DC equations for the converted circuit:

$$\frac{E_{in} - V_1}{R_1} = \frac{V_1}{R_2} + i_{C1}, \quad \text{or} \quad V_1 \left( \frac{1}{R_1} + \frac{1}{R_2} \right) + i_{C1} = \frac{E_{in}}{R_1}$$

And obviously  $V_1 = E_1$ . Forming A1, B2 and P as before:

$$A1 = \begin{bmatrix} \frac{1}{R_1} + \frac{1}{R_2} & 1 \\ 1 & 0 \end{bmatrix}, \quad B2 = \begin{bmatrix} 0 & \frac{E_{in}}{R_1} \\ E1 & 0 \end{bmatrix}, \quad P = C1$$

Using values of  $R_1 = 20K$ ,  $R_2 = 40K$ , and  $C1 = 0.5\mu F$ , calculating V, H, AB, etc., leads to

$$A = -150, \quad B = 100 \quad D = 1 \quad E = 0$$

A is dimension  $\{N \ N\} = \{1 \ 1\}$  as expected.

Since  $\left| \frac{1}{A} \right| = 6.667 \cdot \text{ms}$ , choose  $\Delta t = 500 \cdot \mu\text{s}$ .

We can now change  $E_{in}$  from a 1V input to a 4.5V step input and initialize the capacitor voltage  $V_{c1}$  to zero.  $V_{c1}$  can be set to any initial condition, but zero is usually chosen.

Using the form  $\Delta x = (Ax + Bu)\Delta t$  we have

$$\Delta V_{c1} = (A \cdot V_{c1} + B \cdot E_{in})\Delta t = 0.225$$

$$V_{c2} = \Delta V_{c1} + V_{c1} = 0.225$$

$$\Delta V_{c2} = (A \cdot V_{c2} + B \cdot E_{in})\Delta t = 0.208$$

$$V_{c3} = \Delta V_{c2} + V_{c2} = 0.433$$

Combining:

$$V_{c2} = (A \cdot V_{c1} + B \cdot E_{in}) \Delta t + V_{c1} = 0.225$$

$$V_{c3} = (A \cdot V_{c2} + B \cdot E_{in}) \Delta t + V_{c2} = 0.433$$

Or in general, and setting the maximum number of iterations  $k_{max} = 100$ , and indexing  $k = 2, \dots, k_{max}$

$$V_{c_k} = (A \cdot V_{c_{k-1}} + B \cdot E_{in}) \Delta t + V_{c_{k-1}}, \quad t = k \cdot \Delta t$$

The total time sweep or period is then  $T = k_{max} \cdot \Delta t = 50 \cdot \text{ms}$

The output equation  $y = Dx + Eu$  is

$$V_{o_k} = D \cdot V_{c_k} + E \cdot E_{in} = V_{c_k}$$

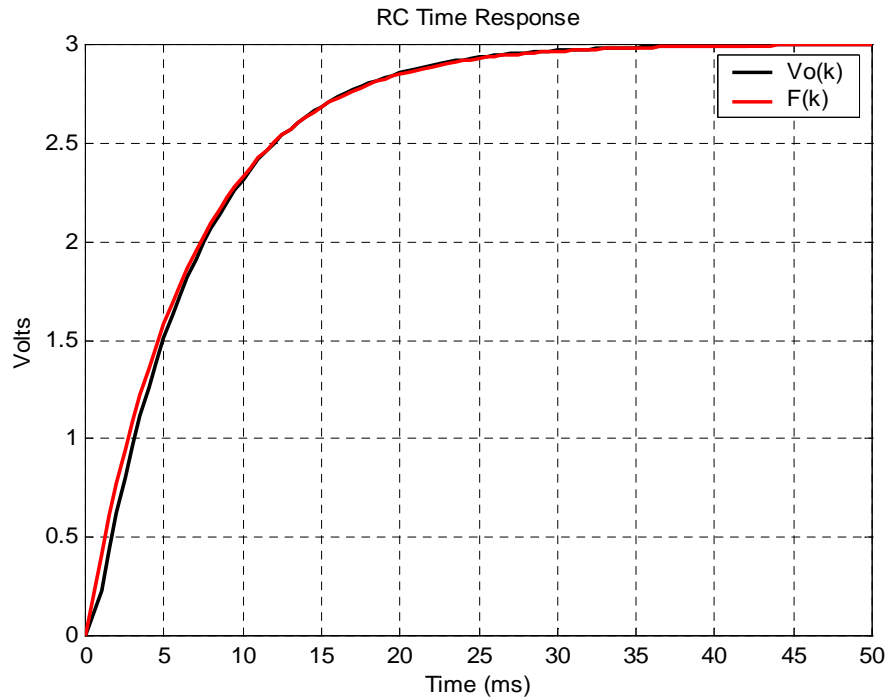
The circuit time constant is

$$\tau = R_p \cdot C_1 = \left| \frac{1}{A} \right| = 6.667 \cdot \text{ms}, \quad 5\tau = 33.33 \cdot \text{ms}, \quad R_p = \frac{R_1 \cdot R_2}{R_1 + R_2}$$

Note that the steady state dc output is  $V_{dc} = \frac{E_{in} \cdot R_2}{R_1 + R_2} = 3V$ , as shown on the time plot.

The output  $V_o(k)$  and  $F(t) = \frac{E_{in} \cdot R_p}{R_1} \left( 1 - \exp\left(\frac{-t}{R_p \cdot C_1}\right) \right)$  is plotted via the M-file `rc_time.m`

As will be seen, this method applies whether A, B, D, & E are scalars or arrays.



Output from TwinT.m

